



POLITECNICO
MILANO 1863

DIPARTIMENTO DI ELETTRONICA
INFORMAZIONE E BIOINGEGNERIA

Multirate processing

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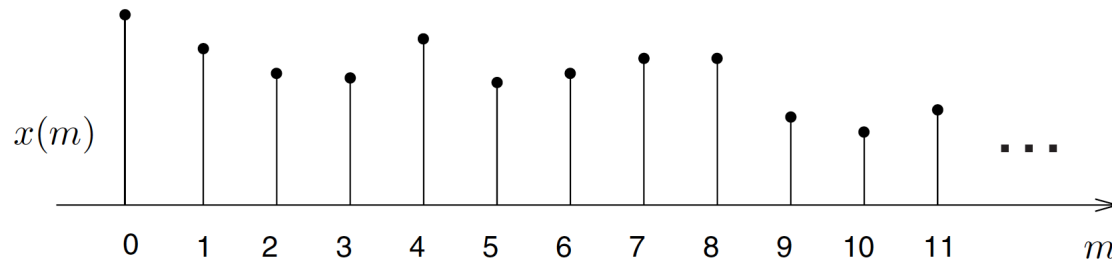
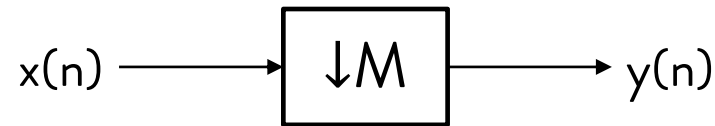
Given a signal $x(n)$, sampled with sampling frequency (or sampling rate) F_s , multirate processing concerns processing the signal with different sampling rate $F_s' \neq F_s$:

- Downsampling and decimation are related to $F_s' < F_s$
- Upsampling and interpolation are related to $F_s' > F_s$

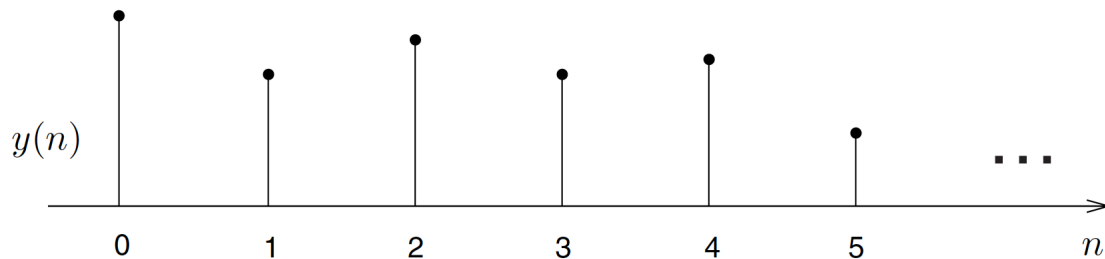
Downsampling

Downsampling of a factor M means to keep one sample every M samples and discard the rest

$$y(n) = x(nM)$$

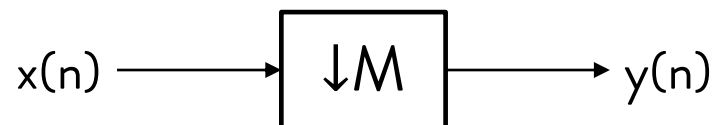


$$M = 2$$



Downsampling

$$y(n) = x(nM)$$



- In Z domain,

$$Y(z) = \sum_{n=-\infty}^{+\infty} y(n)z^{-n} = \sum_{n=-\infty}^{+\infty} x(nM)z^{-n} = \sum_{m=-\infty}^{+\infty} x(m) \left[\frac{1}{M} \sum_{k=0}^{M-1} e^{j\frac{2\pi km}{M}} \right] z^{-\frac{m}{M}}$$

with

$$\frac{1}{M} \sum_{k=0}^{M-1} e^{j\frac{2\pi km}{M}} = \begin{cases} 1 & m \text{ is multiple of } M \\ 0 & \text{otherwise} \end{cases}$$



$$Y(z) = \frac{1}{M} \sum_{k=0}^{M-1} \sum_{m=-\infty}^{+\infty} x(m) \left[e^{-j\frac{2\pi k}{M}} \cdot z^{\frac{1}{M}} \right]^{-m} = \frac{1}{M} \sum_{k=0}^{M-1} X \left(e^{-j\frac{2\pi k}{M}} \cdot z^{\frac{1}{M}} \right)$$

Downsampling

$$y(n) = x(nM)$$



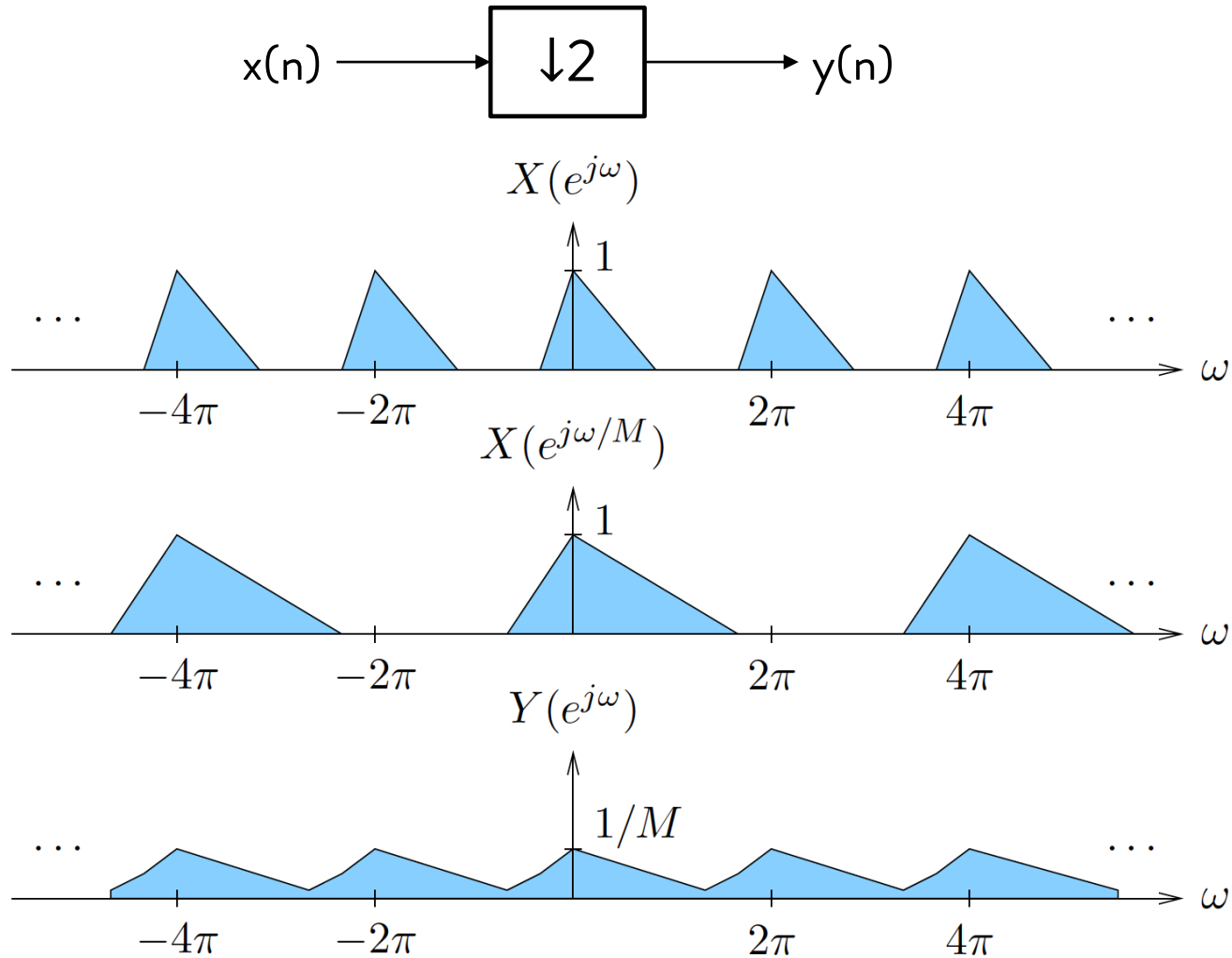
- In frequency domain $Y(z)$ becomes

$$Y(f) = \frac{1}{M} \sum_{k=0}^{M-1} X\left(\frac{f - k}{M}\right)$$



- The DTFT of $y(n)$ is composed of copies of the DTFT of $x(n)$ expanded by M and repeated with period 1 in normalized frequency (or F_s in Hertz, or 2π in angular frequencies)
- The gain is reduced by a factor of M

Downsampling: example

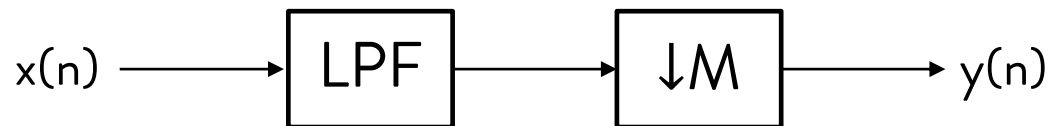


Aliasing in frequency occurs if the DTFT of $x[n]$ is not limited to $1/(2M)$ (or π/M , or $F_s/(2M)$)

Decimation

Decimation is related to downsampling the signal, but avoids frequency aliasing:

- The signal $x(n)$ is filtered with a low-pass filter having cut-off frequency $= 1/2M$
- Then, the filtered signal is downsampled by a factor M



$$H(f) = \begin{cases} 1 & |f| \leq \frac{1}{2M} \\ 0 & \text{otherwise} \end{cases}$$

Ex 25: downsampling and decimation

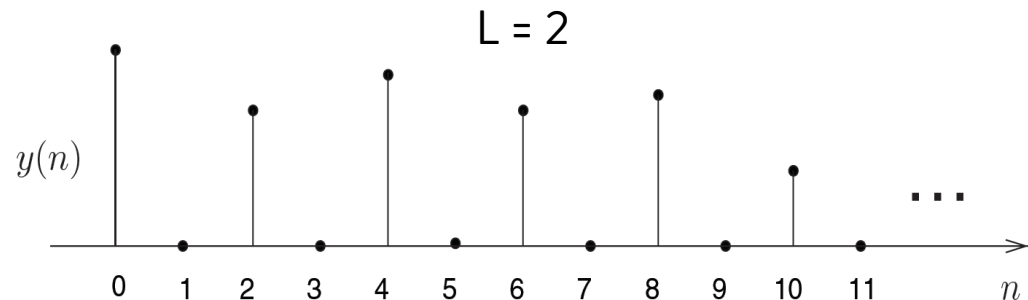
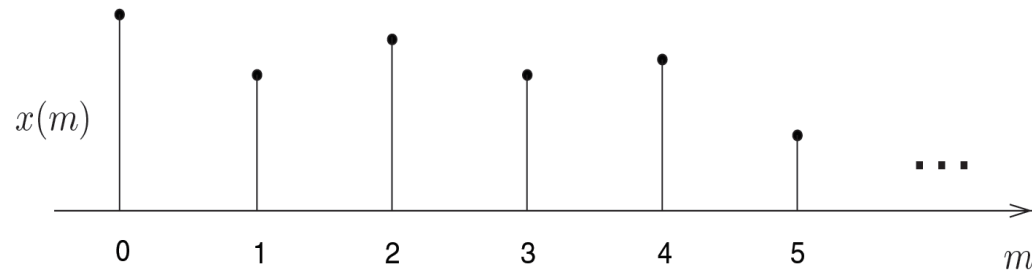
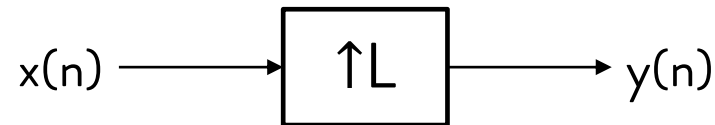
Given $x(n)$ defined as the sum of two sinusoidal signals, sampled at $F_s = 500$ Hz with duration 3 seconds, one with frequency 50 Hz and the other one with frequency 100Hz:

- Downsample $x(n)$ with downsampling factor $M = 4$
- Decimate $x(n)$ with decimation factor $M = 4$, using a FIR filter with order 64.
- Plot the DFTs of $x(n)$, of the downsampled and of the decimated signals vs frequency [Hz] in the same figure and comment on the results.
- Try also $M = 2$ and see what happens

Upsampling

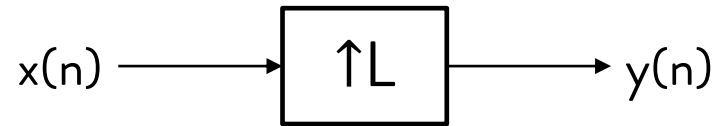
Upsampling of a factor L means to insert $L - 1$ zeros between the input signal samples

$$y(n) = \begin{cases} x(n/L) & n = kL \\ 0 & \text{otherwise} \end{cases}$$



Upsampling

$$y(n) = \begin{cases} x(n/L) & n = kL \\ 0 & \text{otherwise} \end{cases}$$



- In Z domain,

$$Y(z) = \sum_{n=-\infty}^{+\infty} y(n)z^{-n} = \sum_{k=-\infty}^{+\infty} x\left(\frac{kL}{L}\right) z^{-kL} = \sum_{k=-\infty}^{+\infty} x(k)z^{-kL} = X(z^L)$$

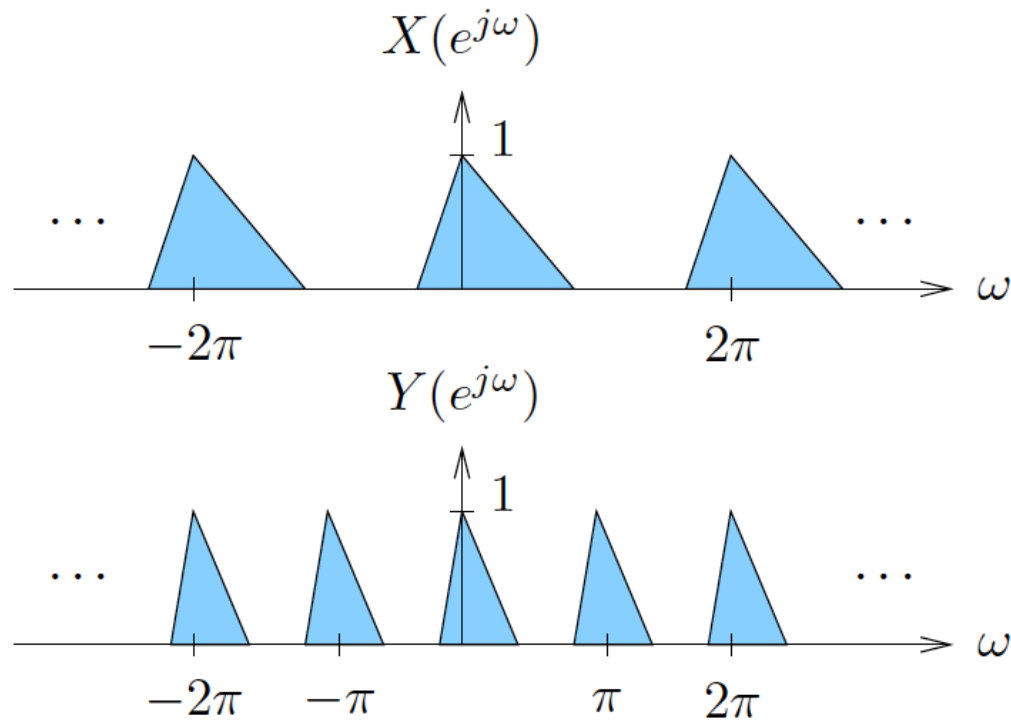
- In frequency domain,

$$Y(f) = X(fL)$$



- Upsampling compresses the DTFT by a factor of L

Upsampling: example

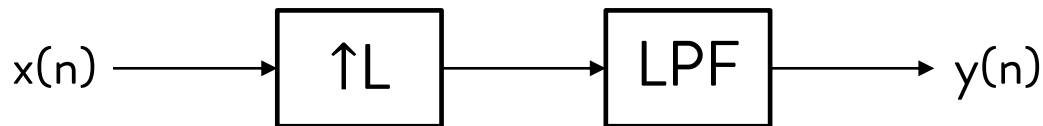


Spectral replicas do not overlap: upsampling just causes a compression of the spectrum, which has a new period of $1/L$ (or $2\pi/L$, or $F_s/(L)$)

Interpolation

Idea: instead of zeros, what if we interpolate signal values?

- First, upsample the signal by a factor L
- Then, filter the signal with a low-pass filter with cut-off $= 1/2L$, which filters out the replicas and interpolate the signal samples



$$H(f) = \begin{cases} L & |f| \leq \frac{1}{2L} \\ 0 & \text{otherwise} \end{cases}$$

Ex 26: upsampling and interpolation

Given the downsampled signal defined in Ex25 with $M = 4$:

- Create the signal x_1 by upsampling the signal with a factor $L = 4$

Given the decimated signal defined in Ex25 with $M = 4$:

- Create the signal x_2 by interpolating the signal with a factor $L = 4$, using a FIR filter with order 64.
- Open a figure and create three subplots:
 1. In 1^o subplot, plot the stem of the original signal $x(n)$ until $N = 200$ time samples, x-axis in seconds.
 2. In 2^o subplot, plot the stem of the downsampled and decimated signals with the same temporal duration as above
 3. In 3^o subplot, plot the stem of x_1 and x_2 with the same temporal duration

Rational sampling rate conversion

Sampling rate change by a factor L/M can be easily implemented by cascading an interpolator with a decimator:



- The low-pass filter is built to delete replicas due to upsampling and avoid frequency aliasing due to downsampling

$$H(f) = \begin{cases} L & |f| \leq \min\{\frac{1}{2L}, \frac{1}{2M}\} \\ 0 & \text{otherwise} \end{cases}$$

Ex 27: rational sampling rate conversion

Given the signal $x(t) = A_1 \cos(2\pi f_1 t) + A_2 \cos(2\pi f_2 t)$:

- Create the signal $x(n)$ as $x(t)$ with t from 0 to 0.5 seconds, sampled at $F_s = 8000$ Hz. $A_1 = 0.7$, $A_2 = 0.5$, $f_1 = 1800$ Hz, $f_2 = 3600$ Hz
- Create the signal $y(n)$ by resampling $x(n)$ with 6000 Hz, without using the MATLAB functions for automatic resampling. Use $N = 64$ filter samples.
- Plot the magnitude of the DFTs of $x(n)$, the upsampled signal, the filtered signal and $y(n)$ over 2048 samples vs normalized frequency in $[0, 1)$