

Ex.1 (Pt.16)

A signal $x(n)$ has been filtered by a linear time-invariant system with impulse response:

$h(n) = \delta(n) - \frac{1}{2}\delta(n-1)$. The output of the system is the distorted signal $y(n)$. Theoretically,

$x(n)$ can be recovered from $y(n)$ through an inverse filter having a system function equal to

the reciprocal of the system function of the distorting filter: $H_i(z) = \frac{1}{H(z)}$.

- 1) Determine the z-transform $H(z)$ of the impulse response, and plot its zeros-poles plot.
- 2) Determine analytically $h_i(n)$ [the generating series], is it a FIR or an IIR filter? Plot its amplitude response.
- 3) If $x(n) = \delta(n) + \delta(n-1) + \delta(n-2)$, calculate $y(n)$ by doing the linear convolution directly.
- 4) Calculate $y(n)$ working directly in the frequency domain using the DFT.
- 5) Determine the z-transform of $Y(z)$ and determine the z-transform of $Y(z) \cdot H_i(z)$.

Ex.2 (Pt.6)

The signal $x(n)$, $x(n) = \{\dots, 0, 0, \underline{1}, 2, 3, 2, 1, 0, 0, \dots\}$, where $\underline{1}$ represents $x(0)$, is applied as the input to the following system:

$$x(n) \longrightarrow \boxed{\uparrow 2} \longrightarrow \boxed{H(z)} \longrightarrow \boxed{\downarrow 3} \longrightarrow y(n)$$

If the impulse response $h(n)$ is given by $h(n) = \{\underline{1}, 2\}$, the what is the output signal $y(n)$?

Using the same impulse response, what will be the output in this case?

$$x(n) \longrightarrow \boxed{\uparrow 2} \longrightarrow \boxed{H(z^2)} \longrightarrow y(n)$$

Ex.3 (Pt. 11 – MATLAB code)

A LTI filter is defined as: $H(z) = \frac{B(z)}{A(z)} = \frac{1 + cz^{-1}}{1 - az^{-1} - bz^{-2}}$

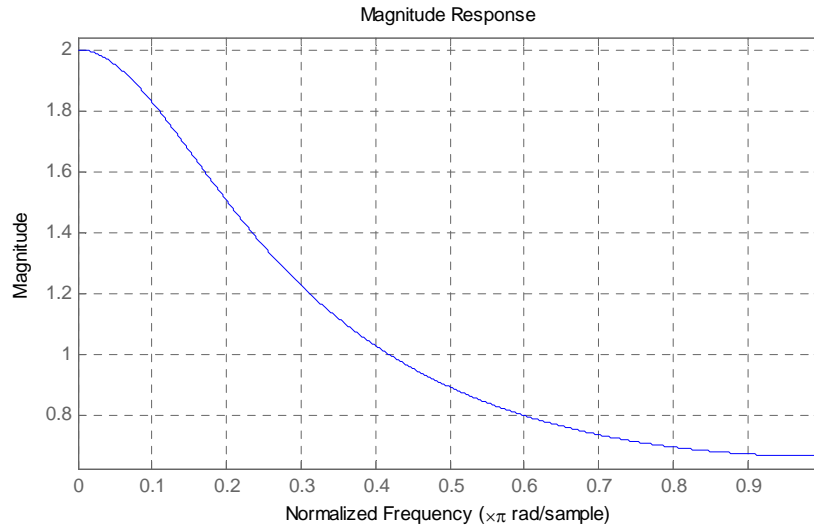
- a) Choose randomly a, b, c, but in order to implement a minimum phase stable filter.
- b) Compute and plot poles and zeros in the complex plane.
- c) Compute and plot the Frequency response.
- d) Compute and plot the impulse response $h(n)$ at $n = 0, \dots, 100$.
- e) Defined $x(n) = \sin(0.1 \pi n)$, compute and plot the output at $n=0, \dots, 100$.

Solutions

Ex.1

$$H(z) = 1 - \frac{1}{2}z^{-1}$$

$$H_i(z) = \frac{1}{1 - \frac{1}{2}z^{-1}}. \text{ It is a IIR stable filter, } h(n) = \left(\frac{1}{2}\right)^n u(n)$$



$$y(n) = \delta(n) + \frac{1}{2}\delta(n-1) + \frac{1}{2}\delta(n-2) - \frac{1}{2}\delta(n-3)$$

$$X(k) = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ 1 \\ 1 \\ 0 \end{bmatrix} = \begin{bmatrix} 3 \\ -j \\ 1 \\ j \end{bmatrix} \quad H(k) = \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & -j & -1 & j \\ 1 & -1 & 1 & -1 \\ 1 & j & -1 & -j \end{bmatrix} \begin{bmatrix} 1 \\ -\frac{1}{2} \\ 0 \\ 0 \end{bmatrix} = \begin{bmatrix} \frac{1}{2} \\ 1 + \frac{1}{2}j \\ \frac{3}{2} \\ 1 - \frac{1}{2}j \end{bmatrix}$$

$$Y(k) = \begin{bmatrix} \frac{3}{2} \\ \frac{1}{2} - j \\ \frac{3}{2} \\ \frac{1}{2} + j \end{bmatrix} \quad y(n) = \frac{1}{4} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & -1 & j \end{bmatrix} \begin{bmatrix} \frac{3}{2} \\ \frac{1}{2} - j \\ \frac{3}{2} \\ \frac{1}{2} + j \end{bmatrix} = \begin{bmatrix} 1 \\ \frac{1}{2} \\ \frac{1}{2} \\ -\frac{1}{2} \end{bmatrix}$$

$$Y(z) = 1 + \frac{1}{2}z^{-1} + \frac{1}{2}z^{-2} - \frac{1}{2}z^{-3}$$

$$Y(z) \cdot H_i(z) = \frac{1 + \frac{1}{2}z^{-1} + \frac{1}{2}z^{-2} - \frac{1}{2}z^{-3}}{1 - \frac{1}{2}z^{-1}}$$

Ex.2

The outputs will be, in the first case: $\{1, 4, 2, 2\}$ while in the second one will be $\{1, 0, 4, 0, 7, 0, 8, 0, 5, 0, 2\}$

Ex.3

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clc
clear all
close all

% y(n) = ay(n-1) + by(n-2) + x(n) + cx(n-1)
% this is a IIR filter , order 2 .
% transfer functionis :
% Y(z)/X(z) = (1 + cz^-1) / (1 - az^-1 - bz^-2)
% this is equal to :
% 1 + cz^-1 / (1 - ( p1 + p2 ) z^-1 + p1 p2 z^-2)
% where p1 and p2 are the poles of the transfer function:
% (z^2 - a z - b )/z^2 = (p1 - z) (p2 - z)/z^2
% (z^2 - a z - b )/z^2 = (p1 z^-1 - 1) (p2 z^-1 - 1)
% (z^2 - a z - b )/z^2 = 1 - ( p1 + p2 ) z^-1 + p1 p2 z^-2)

% Hence we choose |p1| and |p2| < 1 if we want the filter to be stable.
% For instance , p1 = 0.9 and p2 = -0.7 ---> a = 0.2 , b = 0.63.
% Then if we want also a minimum phase filter we choose |c| (the zero)
% inside the unit circle, for example c = 0.8
b = [1 0.8];
a = [1 -0.2 -0.63];
figure, zplane(b,a)

% b
z = roots(b)
p = roots(a)
figure, zplane(z,p)

%without using freqz
%the frequency response is the DTFT of the (truncated) impulse response:
N = 1024;
delta = [1, zeros(1,N-1)];
h = filter(b, a, delta);
figure,
plot(abs(h))

H = fft(h,N);
w = 2*pi*[0 :N-1]/N;
figure,
subplot(2,1,1);
plot (w/pi,20*log10(abs(H))); title('using filter')
axis( [ 0 1 -20 30 ] ) ;
xlabel( 'frequency in pi units');
ylabel( '|H(w)| [ dB ]')
subplot(2,1,2);
plot(w/pi, angle(H)/pi);
axis([ 0 1 -1 1 ]);
xlabel('frequency in pi units');
ylabel('<H(w)in pi units')

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```
% using freqz
figure,
subplot(2,1,1);
[H, w] = freqz(b,a,N);
plot(w/pi,20*log10(abs(H))); title('using freqz')
axis( [ 0 1 -20 30 ] ) ;
xlabel( 'frequency in pi units');
ylabel( '|H(w)| [ dB ]')
subplot(2,1,2);
plot(w/pi, angle(H)/pi);
axis([ 0 1 -1 1 ]);
xlabel('frequency in pi units');
ylabel('<H(w) in pi units')

% filtering
% input
n=[0:1:100];
x= sin((pi/10)*n);
y=filter(b,1,x);
figure,
plot(abs(y))
```